

Journal of Engineering Science and Technology Review 14 (3) (2021) 116 - 123

Research Article

JOURNAL OF Engineering Science and Technology Review

www.jestr.org

A Circularly Polarized Semi-Symmetric Curvature Slot Fractal Antenna for Gain Enhancement

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Received 16 October 2020; Accepted 1 January 2021

Abstract

A Novel Semi-Symmetric curvature slot impregnated Decagonal shaped microstrip patch antenna with miniaturized dimensions of 20 x 12 mm² has been designed and analysed for its working in the frequency range of 4.56 to 12.42 GHz. This paper focuses more on how geometrical changes in the slots affects the impedance, radiation pattern, gain and polarization of the microstrip patch antenna with Defected Ground Structure (DGS). The emphasis is on achieving adequate antenna impedance matching with the help of a nonlinear, semi-symmetric curvature slot to improve the peak gain and to achieve circular polarization. Despite being a single element with wideband characteristics, an average peak gain of 12.96 dB has been recorded for the operating bandwidth of 7.86 GHz. Simulated and measured results of the proposed antenna are analysed w.r.t the Return Loss characteristics pertaining under -10 dB.

Keywords: Semi-Symmetric Curvature Slot, Decagonal Patch, V-Groove in Defected Ground Plane, Non-Linearity, Impedance Matching, Peak Gain & Circular Polarization.

1. Introduction

Upgradation of literature in the field of antennas is cruising with a tremendous pace, as many varieties of antennas are taking shape in the shortest possible intervals. But it is technically proven to be true that, the gain-bandwidth compromise has its own impact and influence on the single element antennas. Therefore, the antenna arrays have become more popular these days for maintaining good gains over the operating bands of frequencies. Imagine if a single element can alone accomplish it! This is of course not out of reach and well on the cards. Geometrical changes in the slots of microstrip patch antennas are very likely to produce the desired results. Let's have a quick review of literature on multiband and wideband antennas with average peak gain as the metric for comparison.

Sanjay Chouhan [1] proposed a Spider-shaped fractal MIMO antenna for WLAN/WiMAX/Wi-Fi/Bluetooth/Cband applications for which the average peak gain achieved is 9.8 dB. Majed O. Dwairi [2] introduced Design and Performance Analysis of Fractal Regular Slotted-Patch Antennas for Ultra-Wideband Communication Systems with average peak gains ranging from 3.5 dB to 4 dB. Mohd Gulman Siddiqui [3] has analysed the Koch-Sierpinski Fractal Microstrip antenna for C/X/Ku-band applications. Inkwinder Singh Bangi [4] published an article on Moore, Minkowski and Koch Curves Based Hybrid Fractal Antenna for Multiband Applications and recorded that the average peak gain obtained over the operating bandwidth is around 7 dB. J. A. Tirado-Mendez [5] has analysed Inscribed Fibonacci Circle Fractal in a Circular Radiator for Ultra-Wideband Antenna Operation and Size Reduction and found

that the average peak gain obtained is 1 dB. R. L. Farias [6] proposed a Dual-Port Dual-Wideband Annular Slot Antenna with Stable Unidirectional Pattern, offering an average peak gain of 8.5 dB.

Aliakbar Dastranj [7] designed A New Compact Circular for Broadband Shape Fractal Antenna Wireless Communication Applications and the average peak gain for the working bandwidth has been recorded as 3.7 dB. Ushaben Keshwala [8] presented Honeycomb shaped Fractal Antenna with Defected Ground Structure for UWB applications, whose average peak gain is 3.11 dB. The proposed Decagonal shaped microstrip patch antenna structure with a Semi-Symmetric closed curvature slot extruded at the centre has produced an impressive average peak gain of 12.96 dB for the operating bandwidth of 7.86 GHz (4.56 to 12.42 GHz), of course with the support of Vshaped groove embedded Defected Ground Structure (DGS).

2.Design and Analysis

The proposed Semi-Symmetric curvature slot microstrip patch antenna has been designed in the HFSS Software, with 20mm x 12mm as the dimensions of FR4 (Glass Epoxy; ϵ_r = 4.4) substrate having a thickness of 1.6mm. Initially, the ground plane dimensions are also chosen as 20mm x 12mm, forming a complete ground plane at the bottom facet of the substrate. A Decagonal patch with a side length of ' α ' = 3.215mm has been designed where, its diameter across five sides (D), perimeter (P) and total area (A) are computed with the help of the following expressions respectively:

$$D = \left(1 + \sqrt{5}\right) * a \tag{1}$$

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(2)

P = 10 * a

$$A = \left(\frac{5}{2}\right) * a^2 * \sqrt{(5 + 2 * \sqrt{5})}$$
(3)

More the discontinuities, more will be the radiating spots on the patch and therefore Decagonal geometry has got some good reason for being chosen in the proposed design. Later, the microstrip feed line is designed with the dimensions 10mm x 3mm and blended with the Decagonal patch, constituting the basic design of the proposed model. The material chosen for the ground plane, microstrip patch and the feed line is PEC (Perfect Electric Conductor). The width of the strip line is optimized based on the following expression.

$$w = \left[\frac{7.48 \times h}{e^{\left(\frac{Z_0\sqrt{\epsilon_r}+1.41}{87}\right)}}\right] - 1.25 \times t \tag{4}$$

Where, \boldsymbol{w} is Width of the Feed-line Z_0 is the Single Ended Impedance \boldsymbol{t} is Trace Thickness \boldsymbol{h} is Substrate Thickness ϵ_r is Relative Dielectric Constant

As an improvement to the design for enhancing the operational characteristics, the length of the ground plane is shortened from 20mm to 8.5mm and an optimized V-shaped groove of side length 4.697mm is introduced as the defected ground structure as shown in the Figure 2. In addition, a semi-symmetric curvature slot of height 9.6mm (optimized) is etched from the patch. The V-shaped groove on the ground plane and the closed curvature slot on the patch in combination has got lot of significance, where the changes in the shape of semi-symmetric curvature slot and depth of V-groove are being reasonably reflected in the characteristics of S11 (Return Loss), Gain, as well as Polarization. This particular design has unveiled the fact that curvature-based slots are much more efficient when compared to the regular polygonal geometry-based slots, in achieving wider bandwidths, as well as enhancing the gains of the microstrip patch antenna and obtaining Circular polarization.

The design of the proposed semi-symmetric curvature slot fractal antenna has been evolved through the iterations starting with the microstrip line, circular patch of radius 5mm, decagonal patch with full ground plane, then the optimized defected ground plane, followed by the inclusion of an optimized V-groove on the ground plane and finally the proposed model as shown in the Figure 1.



Fig. 1. Iteration wise Antenna Design (a) Iteration-0, (b) Iteration-1, (c) Iteration-2, (d) Iteration-3, (e) Iteration-4, (f) Proposed Model



Fig. 2. Proposed Antenna Model

The Semi-Symmetric closed curvature slot in the proposed microstrip patch antenna is a heterogeneous curve (which is a combination of an Inverted Parabola and Neile's Semi-cubical Parabola), whose cartesian equations are expressed as following:

$$y = -bx^{2} + c$$
 (expression for Inverted Parabola) (5)

 $y^3 = mx^2$ (expression for Neile's Semi-cubical Parabola)(6)

Where, m& b are the slopes of Neile's Semi-cubical Parabola and Inverted Parabola respectively, and c is the y intercept.

The semi-symmetric curvature slot on the microstrip patch in line with the V-groove on the ground plane induces non-linearity in the Decagonal patch structure leading to Circular polarization at the resonant frequencies. All the optimized design parameters of the proposed antenna are listed in Table 1.

Table 1. Oj	ptimized l	Design	Parameters	of th	e Pr	oposed	Anten	ina	L

Description of Parameter	Value	Description of Parameter	Value
I enoth of the Substrate (I s)	20 mm	Side length of the Decagon (a)	3.215
Length of the Substrate (Es)	20 11111	Side length of the Decagon (a)	mm
Width of the Substrate (Ws)	12 mm	Height of Semi-Symmetric Curvature Slot (Hssc)	9.6 mm
Length of the Defected Ground Plane (Ldgp)	8.5 mm	Length of the feed line (Lfl)	10 mm
Width of Ground Plane (Wgp)	12 mm	Width of the feed line (Wfl)	3 mm
Depth of V-groove in the DGS (Dvg)	4.25 mm	Height of the Substrate (h)	1.6 mm

The reason behind achieving higher gains for the wide bandwidths with the insertion of optimized semi-symmetric curvature slot is the non-linear nature of the slot, which offers better impedance matching when compared to any other regular or linear geometrical shapes. The nonlinearity in the slot structure enables the exponentially changing impedances of the patch to get matched appropriately and thereby producing fruitful results w.r.t gain, bandwidth and polarization.

3. Results and Discussion

The simulation of designed antenna has been carried out in the HFSS software and various parameters such as Return loss, Gain, Radiation Pattern and Axial Ratio have been analysed. The forefront parameter which outlines the performance of the antenna is the Return loss (in dB). S11 gives the amount of power reflected back to the source in the

form of decibels and the threshold for that is S11 < -10 dB. With reference to this metric, the return loss has been analysed for all the iterations of the design including the final model as shown in Figure 3 (a). The S11 characteristics just got better and better as moved towards the proposed model starting from the iteration-0. These changes are a result of variations in the input impedance of the designed antenna. Inclusion of slots in the patch and in the groundplane caused variations in the input impedance of the antenna and therefore defected ground structure with extruded V-grove has produced better impedance matching and has let the return loss characteristics drift closer to -10 dB, when compared with the former iteration results. Yet, the results were not up to the benchmark. The insertion of Semi-Symmetric curvature slot on the decagonal patch (proposed model) has offered much better impedance matching further and achieved a wideband of 7.86 GHz, from 4.56 12.42 operating to GHz range.



Fig. 3. Return loss characteristics of (a) All the Iterations, (b) Proposed Model (Simulated & Measured)

The S11 result analysis has been performed for different geometrical slots such as pentagonal slot, triangular slot and symmetric curve slot prior to adopting the Semi-Symmetric curvature slot, where none of the former models have produced convincing results. The idea behind the thought of inculcating a semi-symmetric curvature slot is to induce non linearity in the patch, in order to match up with the exponentially varying input impedance of the patch antenna expressed as,

$$Z_{in} = Z_0 \left(\frac{Z_L + jZ_0 \tan(\beta l)}{Z_0 + jZ_L \tan(\beta l)} \right)$$
(7)

This theory has worked out quite well when implemented in the design and the S11 characteristics have suggested that the proposed antenna offers a broad bandwidth, which covers most of the UWB. The simulated and measured Return loss characteristics plot of the proposed model with optimized design parameter values is as shown in the Figure 3(b). The parametric analyses have been carried out for the designed antenna to optimize the results, and identified the best design facets in terms of results. Starting with the length of the ground plane (Ldgp), the parametric analyses have been done for depth of the V-groove in the ground plane (Dvg), width of the feedline (Wfl), as well as the height of semi-symmetric curvature slot (Hssc) as shown in the Figures 4(a), 4(b), 4(c) and 4(d) respectively. The optimized design parameter values are Ldgp = 8.5mm, Dvg = 4.25mm, Wfl = 3mm and Hssc = 9.6mm.

Inclusion of the nonlinear curvature slot on the patch and the V-groove in the ground plane has resulted in circular polarization along with wideband and high peak gains. The exponential nature of the curvature slot has significant contribution in obtaining the circular polarization. Normally the circular polarization exhibited by an antenna can be checked with the help of Axial Ratio (in dB), with the condition AR (dB) < 3dB. The circularly polarized radiation patterns (both simulated and measured) at distinct frequencies are as shown in the Figure 5. The frequencies at which the antenna possesses circular polarization are as given in Table2.

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Table 2. De	etails of frequencies at which circular polarization is a	achieved	
S. No	Frequencies at which Circular Polarization is Achieved (GHz)	Phi (in Degrees)	Theta (in Degrees)
1	8.6, 11.54, 11.68 & 11.86	0 & 90	0
2	6.32, 8.4–8.6, 11.58 & 12.14	0	90
3	7.8-7.95 & 10.55-10.65	90	90
4	5.85-6.05, 8.2 & 10.28	60	30
-10 -15 -20 (BP) -25 -35 -35 -40 -45 -50 -35	4 5 6 7 8 9 10 11 12 13 14 Frequency (GHz)	B Frequency (Dvg = 4.5mm Dvg = 4.5mm 0 = 4.5mm 0 = 4.5mm 0 = 0.5 mm 0 = 0.5 mm
-5 -10 -15 (BP) LS -25 -30 -35 -35 -35 -35 -35 -35 -35 -35 -35 -35	Wfl=2.5mm Wfl=3.5mm Wfl=3.5mm Wfl=3.5mm Wfl=3.5mm Wfl=2.5mm Wfl=3.5mm Wfl=2.5mm Wfl=2.5mm Wfl=2.5mm Wfl=2.5mm Wfl=3.5mm Vfl=2.5mm Vfl=2.5mm Vfl=2.5mm Vfl=3.	10 -5 -10 -15 -20 F5 -25 -30 -35 -40 -45 -50 -3 -40 -45 -50 -5 -5 -5 -7 -7 -7 -7 -7 -7 -7 -7 -7 -7	

Fig. 4. S11 parametric analysis of (a) Length of the defected ground (Ldgp), (b) Depth of the V-groove in DGS (Dvg) (c) Width of the feed line (Wfl) and (d) Height of the semi-symmetric curvature slot (Hssc).

Table 3. Peak Gains and Radiation Efficiencies at distinct frequencies in the operating band

S. No	Frequency (GHz)	Peak or Max. Gain (dB)	Radiation Efficiency (%)
1	4.56	10.39	85.23
2	4.9	8.94	88.91
3	5.2	10.39	93.7
4	5.5	11.28	95.63
5	5.9	11.99	92.4
6	6.42	13.51	86.54
7	8.5	13.58	85
8	10.68	15.60	78.81
9	11.98	17.01	74

With reference to the Return loss characteristics, it is evident that the proposed antenna operates in the range of 4.56 and 12.42 GHz frequencies, ticking for many practical applications such as ISM (5.2 & 5.8 GHz), WLAN (5.2/5.8GHz), WiMAX (5.25–5.85GHz), Wi-Fi (4.9, 5 & 5.9GHz), ITU Radar Bands (C-band (5.25–5.95GHz) and Xband (8.5–10.68GHz)), Miracast (5GHz), Satellite Uplink (C-band (5.925–6.425GHz)) and Satellite Downlink (X-band (11.7–12.2GHz)). The simulated and measured results of circularly polarized Radiation Patterns for the proposed antenna are observed at some of the application frequencies (5.2, 5.9, 8.5 and 11.98 GHz) as shown in the Figure 6. The simulated peak gains (dBs) registered at the selected application frequencies 5.2, 5.9, 8.5 & 11.98 GHz are 10.39, 11.99, 13.58 & 17.01 dBs respectively.

Peak Gain (Max. Gain) and Radiation Efficiency Vs Frequency have been plotted for distinct frequencies in the range of 4.56 to 12.42 GHz as shown in Figure 7. The maximum peak gain obtained is 17.01 dB at 11.98 GHz with a minimum peak gain of 8.94 dB recorded at 4.9 GHz. The

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maximum and the minimum radiation efficiencies achieved are 95.63% and 74% at 5.5GHz and 11.98GHz respectively. The peak gains and radiation efficiencies obtained at various frequencies are as mentioned in Table 3. The surface current densities at the frequencies 5.2, 5.9, 8.5 and 11.98 GHz are shown in Figure 8. The current intensities are concentrated along the axis of the feed line at 5.2 and 5.9 GHz, whereas at 8.5 and 11.98 GHz, the current intensities are focused at the edges of the feed line and the radiating element.



Fig. 5. Simulated and measured results of circularly polarized radiation patterns (LHCP & RHCP respectively) at (a) 5.2 GHz, (b) 5.9 GHz, (c) 8.5 GHz and (d) 11.98 GHz.

The proposed antenna model has been fabricated and tested for compatibility with the simulation results by using a combinational analyser and anechoic chamber, where the measured Radiation Patterns and Return loss characteristics are obtained respectively. The measured results are in close concurrence with the simulated results, where the measured peak gains are recorded as 10.342, 11.994, 12.172 & 14.076 dBs at 5.2, 5.9, 8.5 & 11.98 GHz respectively as shown in the Figure 5(a-d). The fabricated antenna model and the corresponding measured S11 (dB) plot are presented in Figure 9(a) and 9(b) respectively.





Fig. 6. Simulated Radiation Patterns, E-plane and H-plane Patterns (with Co & Cross Polarizations) at (a) 5.2 GHz, (b) 5.9 GHz, (c) 8.5 GHz and (d) 11.98 GHz.



Fig. 7. Peak Gain and Radiation Efficiency Vs Frequency plot.



Fig. 8. Surface current densities at (a) 5.2 GHz (b) 5.9 GHz (c) 8.5 GHz & (d) 11.98 GHz.



Fig. 9. (a) Fabricated Antenna Model & (b) Measured Return loss characteristics of the proposed antenna model

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S.No	Reference	Size ($l \ge mm^2$)	Peak Gain (dB)	Operating Band/s (GHz)
1	[1]	37 x 56	4.8	2.24–2.50, 3.60–3.99, 4.40–4.60 & 5.71– 5.90
2	[2]	36 x 24	6.13	3-8.5 & 10-12
3	[3]	60 x 50	Not Mentioned	5.699, 10.108, 11.100, 15.812 & 17.621
4	[4]	41.5 x 37	20.1	1.67, 2.42, 3.58, 4.55, 7.34, 8.05 & 9.52
5	[5]	100 x 100	1	3.0–14.1
6	[6]	150 x 150	9.75	0.824-0.960 & 1.710-2.165
7	[7]	38 x 36	5	3–21.5
8	[8]	20 x 12	3.11	3.74–13.85
9	Proposed	20 x 12	17	4.56–12.42

 Table 4 Comparison of proposed model with other antenna designs

4. Conclusion

A novel curvature slot impregnated decagonal microstrip patch antenna with defected ground structure with miniaturized dimensions (20 x 12 mm) has been designed and analysed for its performance w.r.t operating bandwidth, gain and polarization. In the process of analysis, it is observed that the proposed antenna with optimized semisymmetric curvature slot on the patch along with the Vgroove in the ground plane have contributed quite significantly in achieving wide operating Bandwidth of 7.86 GHz (from 4.56 to 12.42 GHz), extremely impressive Peak (17 dB) and Average Gains (12.96 dB), as well as, Circular Polarization for the frequencies as mentioned in Table 2. The measured results of the fabricated antenna have shown good agreement with the simulated results and found many applications like ISM, WLAN, Wi-Fi, WiMAX, ITU Radar and Satellite bands. A quick comparison of proposed antenna model with other antenna designs has been displayed in Table 4, which clearly depicts that the proposed antenna model has been the best in terms of size, as well as the gains achieved over wide bandwidth.

Acknowledgements

SR/FST/ET-II/2019/450

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